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(71) Applicants:
• **THE FURUKAWA ELECTRIC CO., LTD.**
Tokyo 100-8322 (JP)
• **Sony Corporation**
Tokyo 141-0001 (JP)

(72) Inventors:
• **Washiro, Takanori,**
The Furukawa Electric Co., Ltd.
Tokyo 100-8322 (JP)
• **Iso, Yoichi,** The Furukawa Electric Co., Ltd.
Tokyo 100-8322 (JP)
• **Imagawa, Toshiyuki**
Shinagawa-ku, Tokyo 141-0001 (JP)
• **Oozeki, Minoru**
Shinagawa-ku, Tokyo 141-0001 (JP)

(74) Representative: **HOFFMANN - EITLE**
Patent- und Rechtsanwälte
Arabellastrasse 4
81925 München (DE)

(54) **Small antenna**

(57) A small antenna (10) characterized by comprising a first meander part (14a) formed in such a manner that a meander conductor travels to a first direction and having a first end and a second end, and a second me-

ander part (14b) formed in such a manner that a meander conductor travels to a second direction different from the first direction and having a first end connected with the second end of the first meander part and a second end.

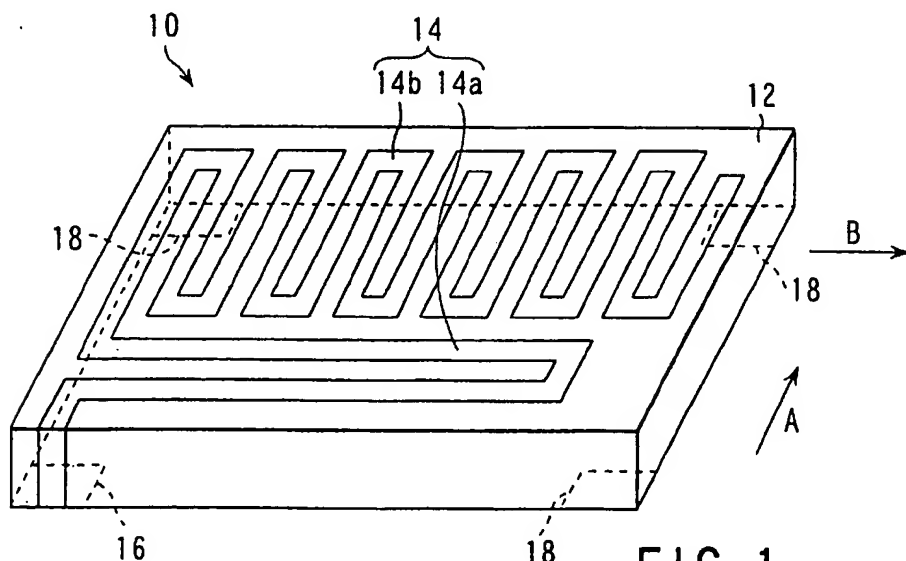


FIG. 1

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Description

[0001] The present invention relates to a small antenna used for a mobile telephone, a mobile information terminal, and a terminal device of a wireless LAN (local area network) etc.

[0002] Conventionally, an antenna in which the antenna conductor is formed on a surface of a dielectric substrate in a meander shape (see Jpn. Pat. Appln. KOKAI Publication No. 10-229304) and the antenna conductor is formed in a helical shape in the dielectric substrate (see Jpn. Pat. Appln. KOKAI Publication No. 10-98322) are well-known as a small antenna which is used for a mobile telephone etc.

[0003] However, when mounting the antenna on a circuit board, it is necessary to mount the antenna directed to a certain direction to show an enough performance as an antenna in a conventional small antenna. Therefore, a conventional small antenna has a small freedom of selection of the mounting direction. Therefore, it is difficult to correspond to the plurality of models with one kind of antenna. Therefore, it takes time of the design, and the cost is raised. In addition, there is a disadvantage that an area necessary for mounting the antenna is enlarged since the conventional antenna should be away from the edge of the ground plate to some degree.

[0004] In the antenna with a meander or helical antenna conductor, by providing the capacity addition part whose width of the conductor is wide to the tip of the antenna conductor (end portion being opposite side of the feeder part), since the length of the antenna conductor can be shortened, it is known that the antenna is miniaturized.

[0005] However, the further miniaturization of an antenna is required in a cellular phone etc.

[0006] An object of the present invention is to provide a small antenna, which has a high degree of freedom when mounting the antenna on a circuit board and is more miniaturized.

[0007] A small antenna according to the present invention is characterized by comprising: a first meander part formed in such a manner that a meander conductor travels to a first direction and having a first end and a second end; and a second meander part formed in such a manner that a meander conductor travels to a second direction different from the first direction and having a first end connected with the second end of the first meander part and a second end.

[0008] This summary of the invention does not necessarily describe all necessary features so that the invention may also be a sub-combination of these described features.

[0009] The invention can be more fully understood from the following detailed description when taken in conjunction with the accompanying drawings, in which:

FIG. 1 is a perspective view of a small antenna according to the first embodiment of the present in-

vention;

FIG. 2A to FIG. 2C are figures showing an example of a method of attaching an antenna to a circuit board of FIG. 1, and FIG. 2A is a plan view, FIG. 2B is a side view and FIG. 2C is a bottom view;

FIG. 3A to FIG. 3C are figures showing another method of attaching an antenna to a circuit board, and FIG. 3A is a plan view, FIG. 3B is a side view, and FIG. 3C is a bottom view;

FIG. 4A to FIG. 4C is a figure showing a still another method of attaching an antenna to a circuit board, and FIG. 4A is a plan view, FIG. 4B is a side view, and FIG. 4C is a bottom view;

FIG. 5 is a perspective view showing a small antenna according to the second embodiment of the present invention;

FIG. 6 is a perspective view showing a small antenna according to the third embodiment of the present invention;

FIG. 7 is a perspective view showing a small antenna according to the fourth embodiment of the present invention;

FIG. 8 is a perspective view showing a small antenna according to the fifth embodiment of the present invention;

FIG. 9 is a perspective view showing a small antenna according to the sixth embodiment of the present invention;

FIG. 10A and FIG. 10B are plan views showing a preferable manner of a capacity addition part provided to a small antenna of the present invention, respectively;

FIG. 11 is a perspective view showing a small antenna according to the seventh embodiment of the present invention;

FIG. 12 is a perspective view showing a small antenna according to the eighth embodiment of the present invention;

FIG. 13A is a plan view of the conventional antenna used in the examination and FIG. 13B is a plan view of a antenna of the present invention;

FIG. 14A and FIG. 14B are graphs showing results of measuring the resonance frequency of the antenna of FIG. 13A and the antenna of FIG. 13B, respectively;

FIG. 15A and FIG. 15B are figures showing the example of an experimental antenna according to the present invention, and FIG. 15A is a plan development view and FIG. 15B is a front view;

FIG. 16A to FIG. 16D are plan views showing a method of attaching an antenna of FIG. 15A and FIG. 15B to the circuit board, respectively; and

FIG. 17A and FIG. 17B are Figure which show the example of an experimental antenna according to the present invention, and FIG. 17A is a plan development view and FIG. 17B is a side view.

[0010] Hereinafter, an embodiment of the present in-

vention will be explained in detail referring to the drawings.

[FIRST EMBODIMENT]

[0011] FIG. 1 is a perspective view of a small antenna according to the first embodiment of the present invention. The small antenna 10 according to the first embodiment has a planar dielectric substrate 12, an antenna conductor 14 provided on a surface of the dielectric substrate 12, and a feeder terminal part 16 provided at a corner portion of another surface of the dielectric substrate 12. This antenna is formed to have substantially a 1/4 wavelength of a transmission/reception frequency signal.

[0012] The antenna conductor 14 has a first meander part 14a and a second meander part 14b. The first meander part 14a is formed in such a manner that the meander conductor travels from the first end (end portion of the feeder terminal part 16 side) arranged at the end portion of the substrate to a certain direction (direction of an arrow A of FIG. 1, that is, short side direction of the substrate). The second meander part 14b is formed in such a manner that the meander conductor travels to a width direction of the meander conductor of the first meander part 14a (direction of an arrow B of FIG. 1, that is, the substrate long side direction) from the second end (end portion of the substrate side is a first end) of the first meander part 14a. It is preferable to lengthen a conductor length of the second meander part 14a more than a conductor length of the first meander part 14b which is connected with the feeder terminal part 16 for the object of widening ratio width. It is also preferable that a meander width of the second meander part 14b is smaller than a meander width of the first meander part 14a for this object. In addition, the first end of the first meander part 14a is connected with the feeder terminal part 16 through the side of the dielectric substrate 12. It is preferable that the pitch of the second meander part 14b is smaller than a size of the meander width of the first meander part 14a and a plurality of pitches are formed to the second meander part 14b along the meander width of the first meander part 14a. The second meander part 14b has about 5.5 pitches in the meander width of the first meander part 14a in this example.

[0013] For this object, it is preferable that the second meander part 14b is extended to an outside of a width direction of the meander conductor of the first meander part 14a. That is, a length size in the pitch direction of the meander conductor of the second meander part 14b becomes larger than a size of the width direction of the meander conductor of the first meander part 14a, and the tip position of the second meander part 14b is arranged on the outside from the end portion in the width direction of the first meander part 14a.

[0014] The fixed terminal parts 18 are provided to a plurality of portions (at three corner portions in the example shown in the figure) away from the feeder terminal

part 16 in the surface of the feeder terminal part 16 on the dielectric substrate 12. This fixed terminal parts 18 are provided to fix the small antenna 10 to the circuit board by soldering.

5 [0015] FIG. 2A to FIG. 4C are figures showing a method of attaching the antenna 10 configured as described above to the circuit board. In FIG. 2A to FIG. 4C, the circuit board 20 comprises an insulation substrate 22. The circuit pattern (omitted in the figure) including a
10 feeder line 24 is formed on one side of the insulation substrate 22, and the ground plate 26 is provided on the other side thereof. The antenna 10 is mounted on the circuit board 20 by soldering the feeder terminal part 16 with the end portion of the feeder line 24, and by soldering the fixed terminal part 18 with the land 28 of the circuit board 20.

[0016] FIG. 2A to FIG. 2C show an example of attaching the antenna to the projection part 20a of the circuit board 20 in such a manner that the long side of the antenna 10 is orthogonal to the edge 26h of the ground plate 26. The ground plate 20 is not provided to the projection part 20a. This attaching method is the same as the method of attaching the conventional meander antenna (whose traveling direction is one direction and is
20 directed to long side direction of the dielectric substrate). Naturally, since the ground plate is influenced hardly by the antenna, this attaching method can show an excellent performance in the small antenna 10 according to the present invention.

[0017] In FIG. 3A to FIG. 3C, a part where the ground plate 26 does not exist in the rear surface of the circuit board 20 is provided. The antenna 10 is attached on the other side of this part in such a manner that long side on the first meander part 14a side is corresponding to the edge 26h of the ground plate 26. When attaching the conventional meander antenna in the direction where the traveling direction of the meander conductor becomes parallel to the edge of the ground plate, the conventional meander antenna cannot show the performance as the antenna when the antenna is not away more than a predetermined distance from the edge 26h of the ground plate 26. According to the antenna 10 of the present invention, the performance as the antenna can be shown sufficiently even in a case of attaching the antenna as shown in FIG. 3A to FIG. 3C. The reason is considered as follows. The traveling direction of the first meander part 14a of the meander conductor is orthogonal to the edge 26h of the ground plate 26. The second meander part 14b becomes a state to be electrically separated from the edge 26h of the ground plate 26 than an actual space by the existence of the first meander part 14a. As shown in FIG. 3A to FIG. 3C, by attaching the antenna 10 so that the long side of the antenna 10 is corresponding to the edge 26h of the ground plate 26, the circuit board 20 can be miniaturized, and the radio set machine can be miniaturized.

[0018] FIG. 4A to FIG. 4C are figures showing an example of providing a notched portion K, in which the

ground plate 26 is cut in the same size as the antenna 10, at one corner portion of the circuit board 20, and attaching the antenna 10 in the opposite side thereof. In the conventional meander antenna, in order to show the performance as an antenna, it is necessary that the notch part of the ground plate is larger than the size of the antenna and the antenna is away from the edge of the notch part of the ground plate, when the antenna is attached in such a manner. In contrast, even if the antenna is attached in such a manner, the antenna 10 according to the present invention can show the sufficient performance as the antenna. It is considered that the reason is similar to the case of FIG. 3A to FIG. 3C. That is, the second meander part 14b becomes a state to be electrically separated in a long distance from the edge K1 along the long side direction of the notch part K of the ground plate 26 by the existence of the first meander part 14a. In addition, the meander conductor travelling direction of the second meander part 14b is orthogonal to the edge Ks along the short side direction of the notch part K of the ground plate 26. The circuit board 20 can be miniaturized and the miniaturization of the radio set machine can be advanced furthermore when the antenna is attached as shown in FIG. 4A to FIG. 4C.

[SECOND EMBODIMENT]

[0019] FIG. 5 is a perspective view of a small antenna according to the second embodiment of the present invention. In FIG. 5, the same mark is fixed to the same part as FIG. 1. In the small antenna 10 according to the second embodiment, the ground terminal part 30 and the feeder terminal part 16 are provided on the surface of dielectric substrate 12, which is opposite to the surface, to which the antenna conductor 14 is provided, and are separated from each other along the meander width direction of the first meander part 14a. The first end of the first meander part 14a is connected with the ground terminal part 30, and the intermediate part is conducted with the feeder terminal part 16.

[0020] The ground terminal part 30 is soldered with the ground conductor of the circuit board, and the feeder terminal part 16 is soldered with the feeder line of the circuit board. The input impedance of the antenna 10 can be adjusted by changing the position connected with the feeder terminal part 16 when the first end of the first meander part 14a is grounded, and power is fed from the intermediate part of the first meander part 14a as mentioned-above. That is, the input impedance lowers when the conduction position with the feeder terminal part 16 is brought close to the ground terminal part 30. The input impedance rises when the branch position of the feeder terminal part 16 is away from the ground terminal part 30. The position is adjusted that the input impedance becomes 50Ω usually.

[THIRD EMBODIMENT]

[0021] FIG. 6 is a perspective view of a small antenna according to the third embodiment of the present invention. In FIG. 6, the same mark is fixed to the same part as FIG. 5. In the small antenna 10 according to the third embodiment, the antenna conductor 14 with the same pattern as that of FIG. 5 is embedded in the dielectric substrate 12. In the third embodiment, a configuration in which the antenna conductor 14 is placed between the dielectric substrates 12 may be applied.

[FOURTH EMBODIMENT]

[0022] FIG. 7 is a perspective view of a small antenna according to the fourth embodiment of the present invention. The fourth embodiment is an embodiment when the present invention is applied to the helical antenna. The small antenna 10 according to the fourth embodiment comprises a rectangular parallelepiped dielectric substrate 12 (showing transparent substrate, for convenience' sake), a helical the antenna conductor 32 embedded in the dielectric substrate 12, a feeder terminal part 16 provided at one corner portion on the bottom of the dielectric substrate 12.

[0023] The antenna conductor 32 has a first helical part 32a and a second helical part 32b. The first helical part 32a is formed in such a manner that the meander helical travels from the first end of the feeder terminal part 16 side to a certain direction (direction of an arrow A of FIG. 1, that is, short side direction of the substrate). The second helical part 32b is formed in such a manner that the helical conductor travels to the helical long diameter direction of the first helical part 32a from the second end of the first helical part 32a (direction of an arrow B, that is, the substrate long side direction). It is preferable that a conductor length of the second helical part 32b is longer than a conductor length of the first helical part 32a for the object of widening the ratio band. It is preferable that a helical diameter of the second helical part 32b is smaller than a helical diameter of the first helical part 32a. The first end of the first helical part 32a is connected with the feeder terminal part 16 through the side surface of the dielectric substrate 12. A pitch of the second helical part 32b is smaller than a size of the helical long diameter of the first helical part 32a, and a plurality of pitches of the second helical part 32b are formed within the range of the helical long diameter of the first helical part 32a.

[0024] The fixed terminal parts 18 are provided to a plurality of portions (to three corner portions in the example shown in the figure) away from the feeder terminal part 16 on the surface of the feeder terminal part 16 side of the dielectric substrate 12. The fixed terminal part 18 is used to fix the small antenna 10 to the circuit board by soldering etc.

[0025] The antenna according to the fourth embodiment can be used similar to the antenna of the first em-

bodiment. The input impedance of the antenna can be adjusted similar to the second embodiment if the first end of the first helical part 32a is connected to the ground terminal part and the intermediate part is connected to the feeder terminal part.

[FIFTH EMBODIMENT]

[0026] FIG. 8 is a perspective view of a small antenna according to the fifth embodiment of the present invention. In FIG. 8, the same mark is fixed to the same part as FIG. 1. The antenna 10 comprises a meander antenna conductor 14 provided on an upper surface of a planar dielectric substrate 12, a capacity addition part 14c which is provided continuously on the second edge of antenna conductor 14 and has a wide conductor width, and a feeder terminal part 16 provided on an under surface of the dielectric substrate 12 on the first edge side of the antenna conductor 14. The first end of the antenna conductor 14 is connected with the feeder terminal part 16 through the side surface of the dielectric substrate 12. The point, of which the fifth embodiment is different from the first embodiment, is only to comprise the capacity addition part 14c, and a detailed explanation will be omitted. By providing the capacity addition part 14c, it is well-known to be able to shorten the conductor length of the antenna conductor 14, but a synergy effect can be obtained when the meander parts 14a, 14b with different meander direction, and the capacity addition part 14c are combined. That is, an antenna formed by forming the antenna conductor 14 having a plurality of meander parts 14a and 14b whose traveling directions of the meander conductors are different and connecting the capacity addition part 14c thereto is compared with an antenna formed by connecting the capacity addition part to the meander conductor whose traveling direction is one direction. If the length of the antenna conductor is the same, the resonance frequency of the antenna having the plurality of meander parts 14a and 14b whose traveling directions of the meander conductor are different is low. In other words, if the target resonance frequency is constant, the antenna conductor can be shortened and the antenna can be miniaturized. If the size of the antenna is assumed to be the same, since the antenna conductor 14 can be shortened, the pitch can be enlarged, the conductor interval can be widened and the bandwidth can be widened.

[0027] The reason is considered as follows.

[0028] Even if the capacity addition part is connected to the antenna conductor whose traveling direction of the meander conductor is one direction, there is a tendency to which the effect, which lowers the resonance frequency, becomes small when the meander frequency increases. But it has been understood that the resonance frequency is effectively lowered, when the traveling direction of the meander conductor is changed on the way even if the meander frequency increases. Therefore, if the antenna conductor is configured by the

plurality of meander parts whose traveling directions of the meander conductors are different, the resonance frequency can be lowered than the case that the traveling direction of the meander conductor is one direction. It becomes possible to miniaturize the antenna.

[SIXTH EMBODIMENT]

[0029] FIG. 9 is a perspective view of a small antenna according to the sixth embodiment of the present invention. In FIG. 9, the same mark is fixed to the same part as FIG. 8. The point of which a small antenna according to the sixth embodiment is different from a small antenna of FIG. 8 is an undermentioned point.

(1) The capacity addition part 14c has a triangle shape in which the conductor width becomes widened by being away from the tip part of the second meander part 14b.

(2) The ground terminal part 30 and the feeder terminal part 16 are provided under the dielectric substrate 12 along the width direction of the meander conductor of the first meander part 14a and are separated with each other similar to the second embodiment (Refer to FIG. 5). The first end of the first meander part 14a is connected with the ground terminal part 30, and the intermediate part is conducted to the feeder terminal part 16.

[0030] The bandwidth can be widened when the capacity addition part 14c is formed to a triangle as shown in FIG. 9. Not only the triangle as shown FIG. 9 but also various shapes can be applied as a shape of the capacity addition part 14c. For example, a pyramid-shape to which the width of the conductor extends in stages as FIG. 10A may be acceptable. A T-shape in which the width of the conductor is widened abruptly at a position which is away from the tip part in the second meander part 14b like FIG. 10B may be acceptable. Thus, the effect which widens the bandwidth can be achieved by widening the tip part of the capacity addition part 14c.

[SEVENTH EMBODIMENT]

[0031] FIG. 11 is a perspective view of a small antenna according to the seventh embodiment of the present invention. In FIG. 11, the same mark is fixed to the same part as FIG. 9. A small antenna according to the seventh embodiment, the antenna conductor 14 and the capacity addition part 14c are embedded similar to the third embodiment in the dielectric substrate 12.

[EIGHTH EMBODIMENT]

[0032] FIG. 12 is a perspective view of a small antenna according to the eighth embodiment of the present invention. In FIG. 12, the same mark is fixed to the same part as FIG. 1 and FIG. 7. The third embodiment is an

embodiment in which the present invention is applied to the helical antenna similar to the fourth embodiment. A small antenna according to the eighth embodiment can be also used similar to a small antenna according to the fourth embodiment.

[0033] A case of which the antenna conductor is configured by two meander parts (the first meander part and the second meander part) whose traveling directions of the meander conductors are different is explained in each above-mentioned embodiment. The present invention is not limited to this, the antenna may have three or more meander parts whose traveling directions of the meanders are different (for example, the third meander part whose traveling direction of the meander conductor is different from an antenna of which the second meander part is provided at the tip part in the second meander part in etc.). In short, in the present invention, the antenna conductor may be configured only by the plurality of meander parts whose traveling directions of the meander conductors are different. It is similar to the helical the antenna conductor.

[EXPERIMENTAL EXAMPLE 1]

[0034] First, to verify the effect by the shape of the antenna conductor according to the present invention, the antenna as shown in FIG. 13A and FIG. 13B is made for trial purposes. FIG. 13A is a conventional antenna whose traveling direction of the meander conductor of the antenna conductor is only one direction. FIG. 13B is an antenna according to the present invention whose traveling direction of the meander conductor of the antenna conductor is two directions. Both antennas have a meander part of the conductor length = 30 mm, line width = 0.2 mm and line interval = 0.2 mm, and the capacity addition part of two equal size triangle of base = 2.2 mm and height = 3 mm is attached in the tip part thereof.

[0035] FIG. 14A is a result of which the resonance frequency of the antenna of FIG. 13A is measured, and FIG. 14B is a result of measuring the resonance frequency of the antenna of FIG. 13B. According to the result, the resonance frequency of the conventional antenna shown in FIG. 13A is 3.01 GHz, and the resonance frequency of the antenna according to the present invention shown in FIG. 13B is 2.66 GHz. Therefore, it can be understood that the resonance frequency of the antenna according to the present invention becomes lower than the conventional one by a large amount even if the sizes thereof are the same. Therefore, if it is the same resonance frequency, the antenna of the present invention can be miniaturized.

[EXPERIMENTAL EXAMPLE 2]

[0036] Next, the antennas as shown in FIG. 15A and FIG. 15B are made for trial purposes. A pattern of the antenna conductor 14 is similar to the embodiment of

FIG. 9. A point different from the embodiment of FIG. 9 is that two fixed terminal parts 18 are formed to connect with the second meander part 14b and the capacity addition part 14c. These terminal parts are actually folded the bottom side of the dielectric substrate 12 as shown in FIG. 15B though the feeder terminal part 16, the ground terminal part 30, and the fixed terminal part 18 are shown in a shape to be developed in FIG. 15A.

[0037] This antenna made for trial purposes is for 2.45 GHz band bluetooth and has a size (size of the dielectric substrate 12) of $8 \times 3 \times 0.4$ (mm). The conductor width of the antenna conductor 14 and the conductor interval are 0.2 (mm). The material of the dielectric substrate 12 is ceramics plastic compound material with the permittivity of 20.

[0038] The antenna 10 made for trial purposes is mounted on the circuit board in such a manner that the position with the ground plate may become FIG. 16A to FIG. 16D, and the performance of the antenna is measured. Table 1 shows the result.

Table 1

| ATTACHING METHOD | BANDWIDTH (MHz) |
|------------------|-----------------|
| FIG. 16A | 290 |
| FIG. 16B | 239 |
| FIG. 16C | 115 |
| FIG. 16D | 124 |

[0039] The bandwidth of 83.5 MHz or more is requested to the antenna for 2.45 GHz band bluetooth, but according to Table 1, it is clear to satisfy this request enough even if the antenna of the present invention are attached by various scheme as shown in FIG. 16A to FIG. 16D. The bandwidth is defined as the range of the frequency which satisfies the relationship of $VSWR < 2$.

[0040] In the conventional antenna, when the antenna is attached to the edge of the ground plate 26 from the side, for example, as shown in FIG. 16C, according to Yujiro Dakeya et al "Chip Multilayer Antenna for 2.45 GHz-Band Application Using LTCC Technology" 2000, IEEE MTT-S International Microwave Symposium Digest (Boston Massachusetts 11-16 June 2000), it is necessary to attach the antenna by separating it from the edge of the ground plate by about 3 mm or more to obtain the bandwidth of 83.5 MHz or more. In the antenna of the present invention, the bandwidth of 115 MHz can be obtained even when the distance from the end of the ground plate is 0.

[0041] In the conventional antenna, when the antenna is attached to the circuit substrate not to project the antenna from the corner portion thereof, for example, as shown in FIG. 16D, it is preferable that the size of the notch part of the corner portion of the ground plate is assumed that the distance between the short side of the notch part and the antenna is 2 mm or more and the

long side of the notch part and the antenna is 5 mm or more according to the Jpn. Pat. Appln. KOKAI Publication No. 10-229304. In contrast, the antenna of the present invention can show the sufficient performance as the antenna even if the distance of the edge of the notch part of the ground plate and the antenna is 0 (even if the size of the notch part of the ground plate is the same as that of the antenna).

[EXPERIMENTAL EXAMPLE 3]

[0042] The antenna which attached the second meander part 14b on the second end of the first meander part 14a side of the antenna conductor 14 is made for trial purposes as shown in FIG. 17A and FIG. 17B. This antenna is formed to have substantially a 1/4 wavelength of a transmission/reception frequency signal. A point different from a small antenna of FIG. 15A and FIG. 15B is as follows.

(1) The extension part 14d is connected to the first end of the first meander 14a of the antenna conductor 14 and is bent in a direction orthogonal to the first meander part 14a (direction of the pitch of the first meander part 14a) in an L-shape. And, the first end of the first meander 14a of the antenna conductor 14 is extended on the side where the second meander part 14b is arranged.

(2) Two fixed terminals 18 are formed to connect with the first meander part 14a and capacity addition part 14c and the terminal parts 16, 18, and 30 are bent to outside in same plane as the bottom of the dielectric substrate 12 as show in FIG. 17B.

(3) The capacity addition part 14c is formed in a rectangle shape. Even if the capacity addition part 14c is a rectangle like this, since the second meander part 14b is extended outside of the width direction of meander conductor of the first meander part 14a, the second meander part 14b can be connected with a center of the capacity addition part 14c and the function as capacity addition part 14c can be properly shown.

[0043] Even if the antenna manufactured as mentioned above is attached in various manners as shown in FIG. 16A to FIG. 16D, the performance as the antenna can be shown enough. Especially, when the antenna is arranged in the notch part K of the ground plate 26 as shown in two-dot chain line in FIG. 17A, it is expected that the influence of the ground plate 26 can be decreased furthermore, and substrate 22 can be further miniaturized. That is, when the traveling direction of the meander conductor of the meander antenna is parallel to the edge of the ground plate 26 as mentioned above, the distance from the end of the ground plate 26 should be made large in general. In the antenna according to this experimental example, the influence of the ground plate 26 to the second meander part 14b is buffered by

the first meander part 14a and the influence of the ground plate 26 to the first meander part 14a it is buffered by the extension part 14d, therefore the performance can be sufficiently shown as an antenna, even if the distance from end K1 and Ks of the ground plate 26 is shortened.

[0044] In the example of the antenna, the terminal parts 18 and 30 may be use as the feeder terminal.

[0045] As described above, the small antenna according to present invention is characterized by comprising: a first meander part formed in such a manner that a meander conductor travels to a first direction and having a first end and a second end; and a second meander part formed in such a manner that a meander conductor travels to a second direction different from the first direction and having a first end connected with the second end of the first meander part and a second end. With this configuration, it is preferable to comprise a feeder terminal part with which the first end of the first meander part is connected and is preferable to comprise a ground terminal part with which the first end of the first meander part is connected; and a feeder terminal part with which an intermediate part in the first meander part is connected.

[0046] Another small antenna according to present invention is characterized by comprising: a meander antenna conductor; and a capacity addition part whose conductor width is wide, provided to a second end of the antenna conductor, and the antenna conductor comprises a plurality of meander parts whose traveling directions are different.

[0047] Another small antenna according to the present invention is characterized by comprising: a first helical part formed in such a manner that a helical conductor travels to a first direction and having a first end and a second end; and a second helical part connected with the second end of the first helical part, formed in such a manner that a helical conductor travels to a direction different from the first direction, and having a first end connected with the second end of the first meander part and a second end. With this configuration, it is preferable to comprise a feeder terminal part with which the first end of the first helical part is connected and is preferable to comprise a ground terminal part with which the first end of the first helical part is connected; and a feeder terminal part with which an intermediate part in the first helical part is connected.

[0048] Another small antenna according to the present invention is characterized by comprising: a meander antenna conductor; and a capacity addition part whose conductor width is wide, provided to a second end of the antenna conductor, and the antenna conductor comprises a plurality of meander parts whose traveling directions are different.

[0049] In each of above small antennas, the following manners are preferable. The following manners are applied solely or by combining them properly.

(1) The antenna conductor (including first meander part and second meander part) is provided on the surface of the dielectric substrate or in the dielectric substrate.

(2) The first meander part (helical part) and the second meander part (helical part) are orthogonal.

(3) The conductor length of the second meander part (helical part) is longer than the conductor length of the first meander part (helical part).

(4) The meander width (helical width) of the second meander part (helical part) is smaller than the meander width (helical width) of the first meander part (helical part).

(5) The pitch (helical pitch) of the second meander part (helical part) is smaller than the meander width (helical width) of the first meander part (helical part).

(6) A plurality of pitches of the second meander part (helical part) are formed within the meander width (helical width) of the first meander part (helical part).

[0050] As mentioned above, according to the present invention, it is possible to correspond to the plurality kinds of models with only one antenna, since the degree of freedom in the direction of the antenna to the ground plate is enlarged when the antenna is mounted on the circuit board. Therefore, a mass production is improved, and the cost reduction can be achieved. Since the antenna can be arranged close to the edge of the ground plate, it becomes possible to reduce an area necessary for mounting the antenna and it is valid in the miniaturization of the radio set machine.

[0051] As explained above, according to the present invention, the meander antenna conductor or the state of helical is configured by the plurality of meander parts or the plurality of helical parts whose traveling directions of the meander conductors (helical conductors) are different. Therefore, since the resonance frequency can be lowered, the length of the antenna conductor can be shortened as a result, and a small antenna having the capacity addition part can be further miniaturized.

Claims

1. A small antenna **characterized by** comprising:

a first meander part (14a) formed in such a manner that a meander conductor travels to a first direction and having a first end and a second end; and

a second meander part (14b) formed in such a manner that a meander conductor travels to a second direction different from the first direction and having a first end connected with said second end of said first meander part (14a) and a second end.

2. The small antenna according to claim 1, **character-**

ized by further comprising a feeder terminal part (16) with which said first end of said first meander part (14a) is connected.

3. The small antenna according to claim 1, **characterized by** further comprising:

a ground terminal part (30) with which said first end of said first meander part (14a) is connected; and
a feeder terminal part (16) with which an intermediate part in said first meander part (14a) is connected.

4. A small antenna **characterized by** comprising:

a first helical part (32a) formed in such a manner that a helical conductor travels to a first direction and having a first end and a second end; and
a second helical part (32b) connected with the second end of said first helical part (32a), formed in such a manner that a helical conductor travels to a direction different from the first direction, and having a first end connected with said second end of said first meander part (14a) and a second end.

5. The small antenna according to claim 4, **characterized by** further comprising a feeder terminal part (16) with which said first end of said first helical part (32a) is connected.

6. The small antenna according to claim 4, **characterized by** further comprising:

a ground terminal part (30) with which said first end of said first helical part (32a) is connected; and
a feeder terminal part (16) with which an intermediate part in said first helical part (32a) is connected.

7. A small antenna **characterized by** comprising:

a meander antenna conductor (14); and

a capacity addition part (14) whose conductor width is wide, provided to a second end of the said antenna conductor, wherein said antenna conductor (14) comprises a plurality of meander parts (14a, 14b) whose traveling directions are different.

8. A small antenna **characterized by** comprising:

a helical antenna conductor (32); and
a capacity addition part (14c) whose conductor

width is wide, provided to a second end of the said antenna conductor, wherein said antenna conductor comprises a plurality of helical parts whose traveling directions are different.

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9. The small antenna according to any one of claim 1, claim 2, claim 3 or claim 7, **characterized in that**

a meander width of said second meander part (14b) is smaller than a meander width of said first meander part (14a),
a pitch of said second meander part (14b) is smaller than a meander width of said first meander part (14a), and
a plurality of pitches of said second meander part (14b) are formed in the meander width of said first meander part (14a).

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10. The small antenna according to any one of claim 4, claim 5, claim 6 or claim 8, **characterized in that**

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a helical width of said second helical part (32b) is smaller than a helical width of said first helical part (32a),
a helical pitch of said second helical part (32b) is smaller than the helical width of said first helical part (32a), and
a plurality of pitches of said second helical part (32b) are formed in the helical width of said first helical part (32a).

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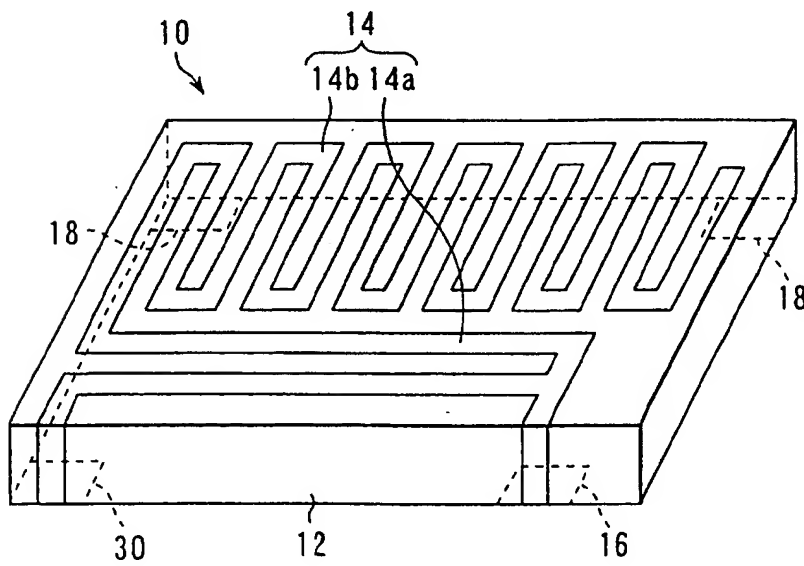
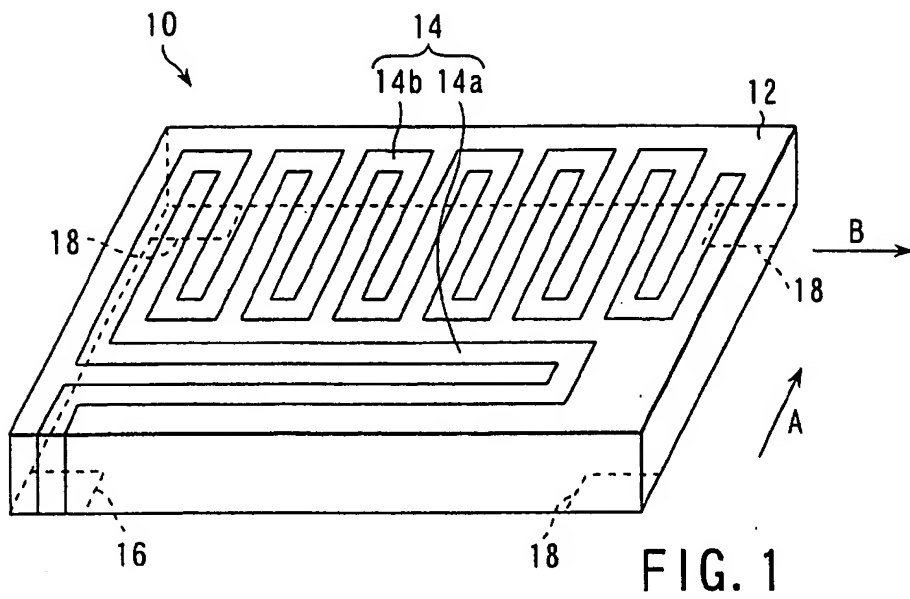
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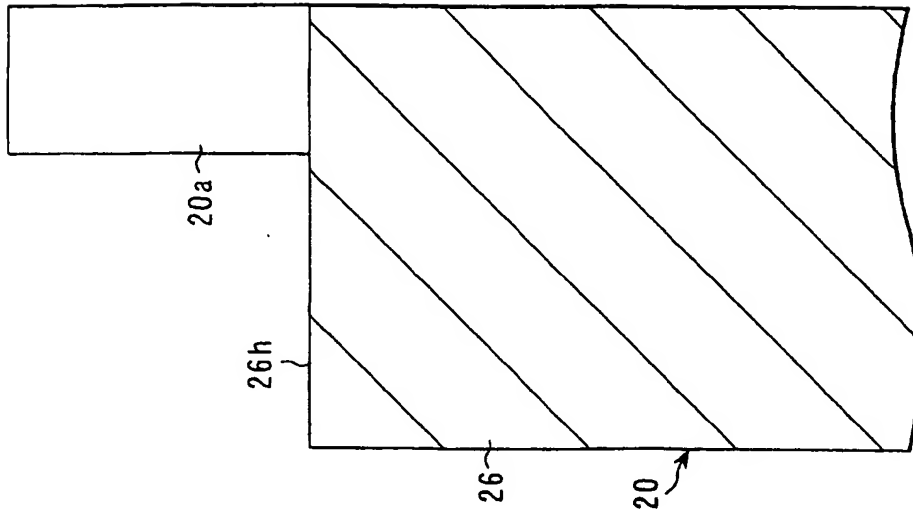


FIG. 2C

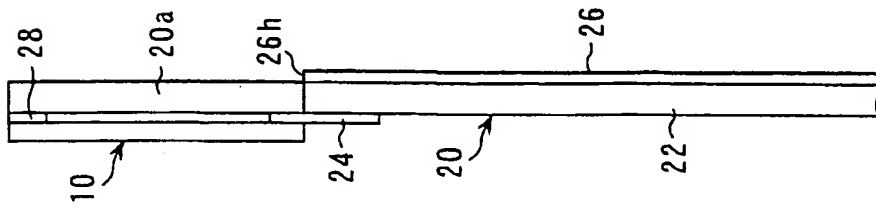


FIG. 2B

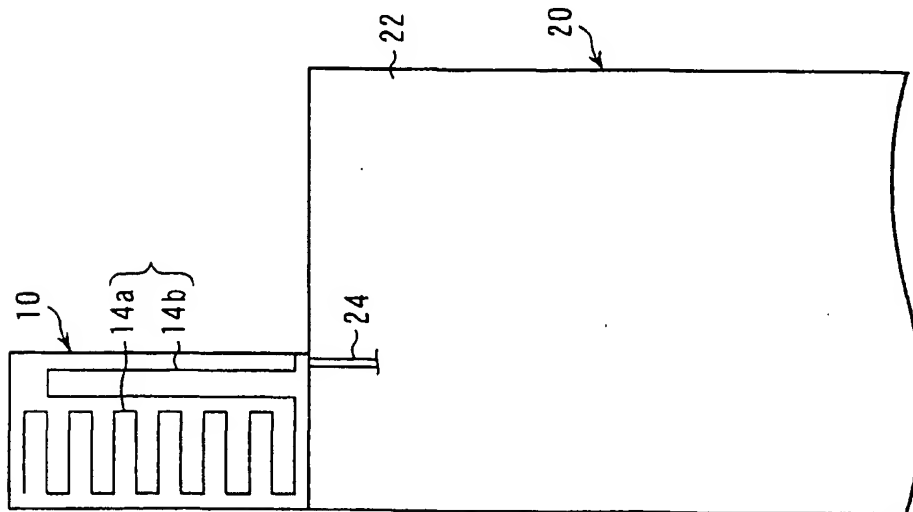


FIG. 2A

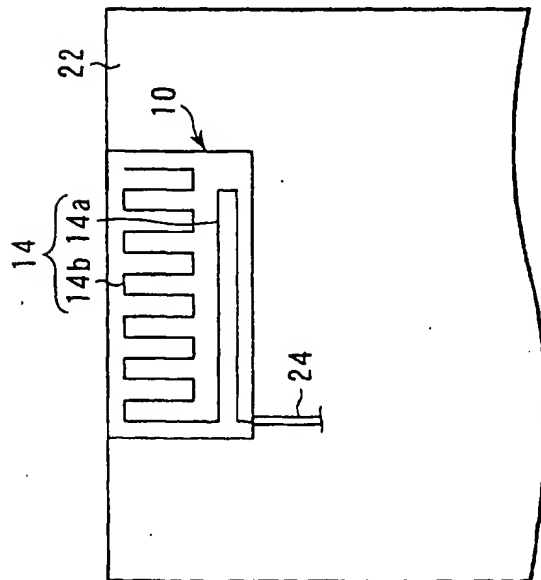


FIG. 3A

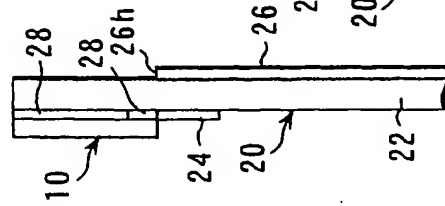


FIG. 3B

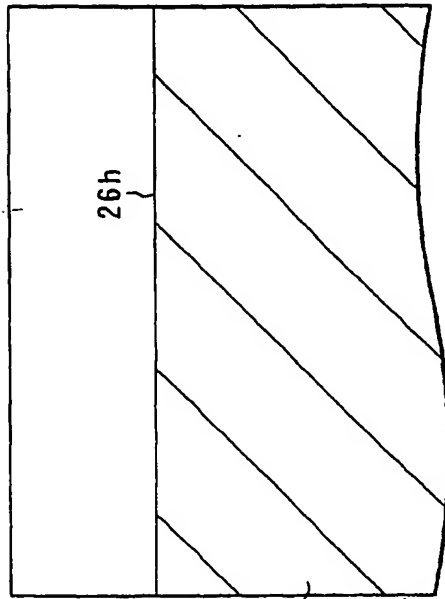


FIG. 3C

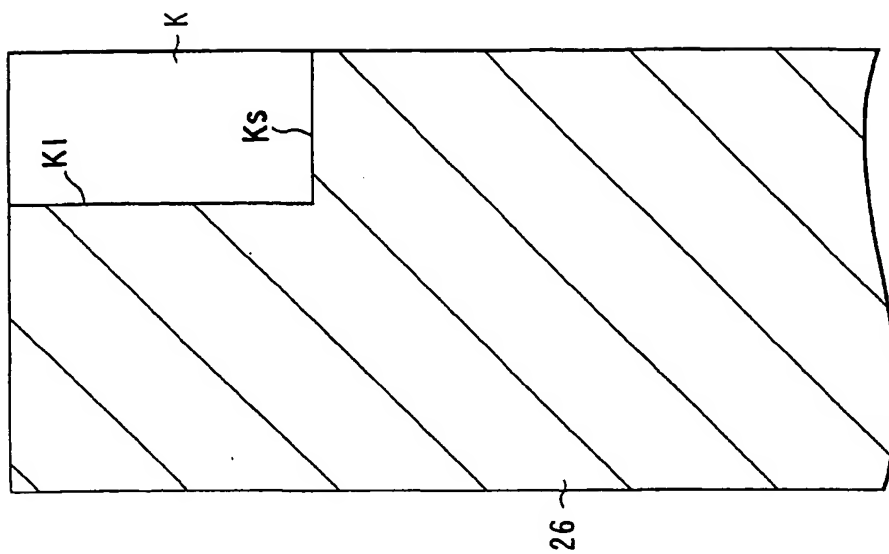


FIG. 4C

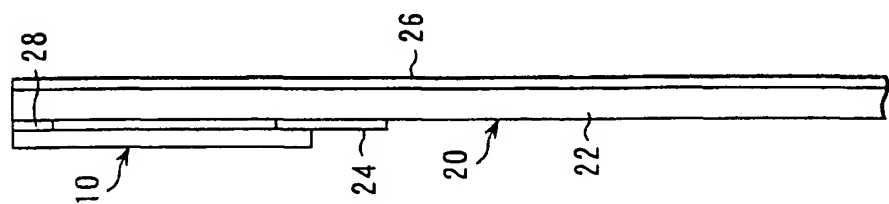


FIG. 4B

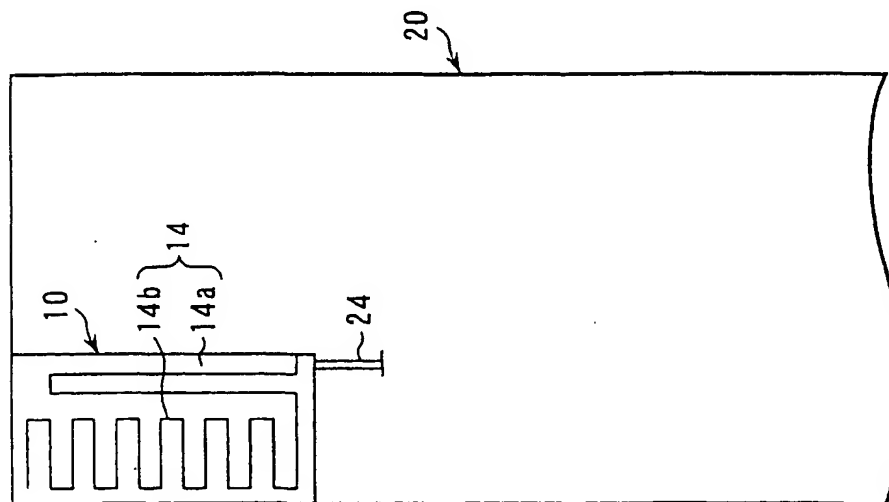
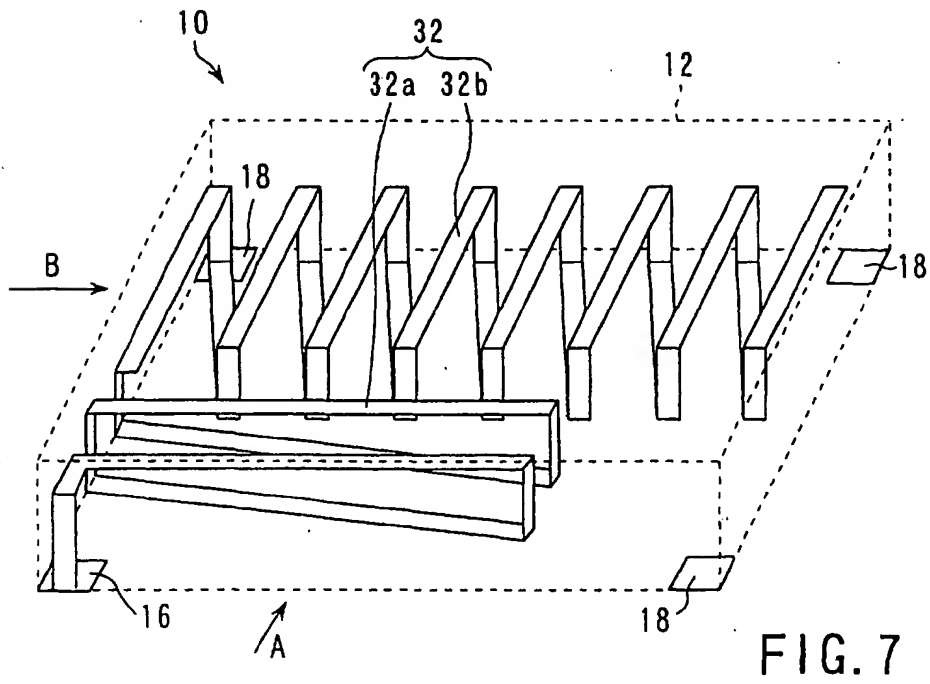
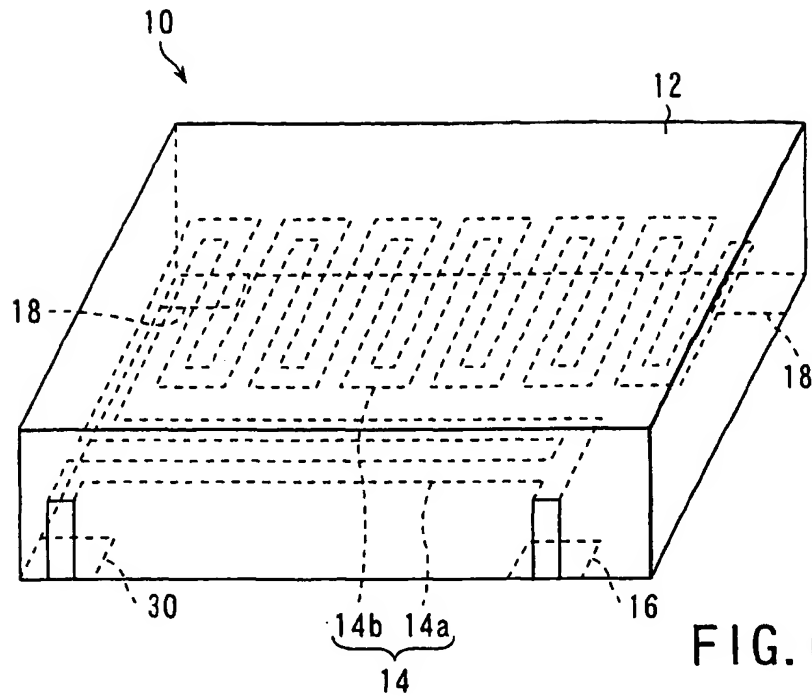
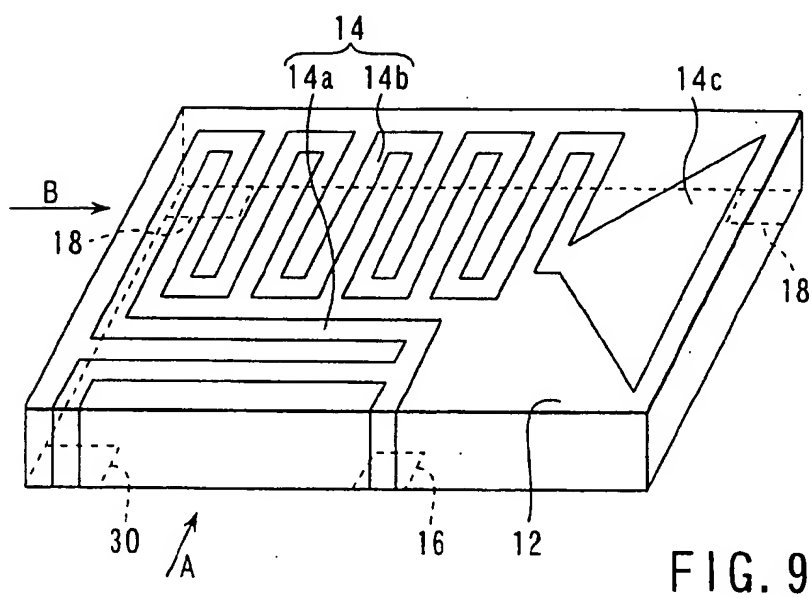
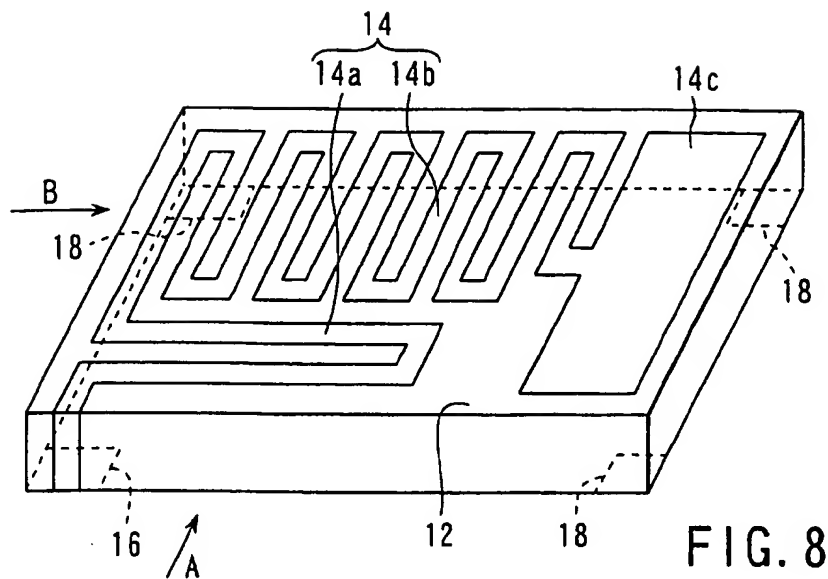


FIG. 4A





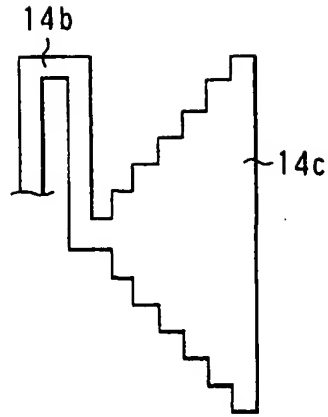


FIG. 10A

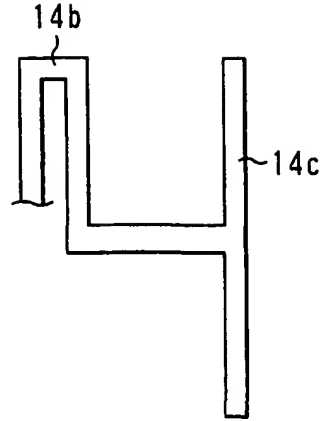


FIG. 10B

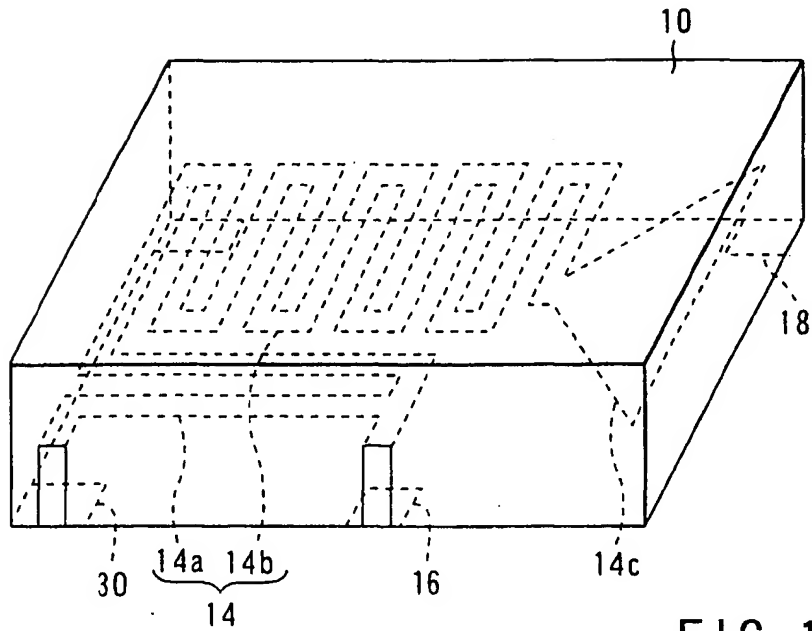
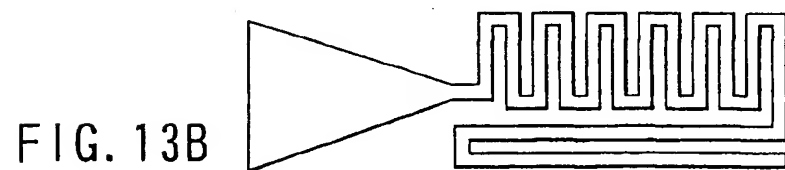
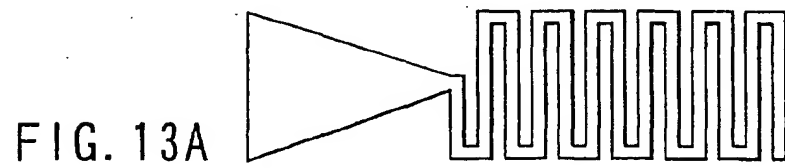
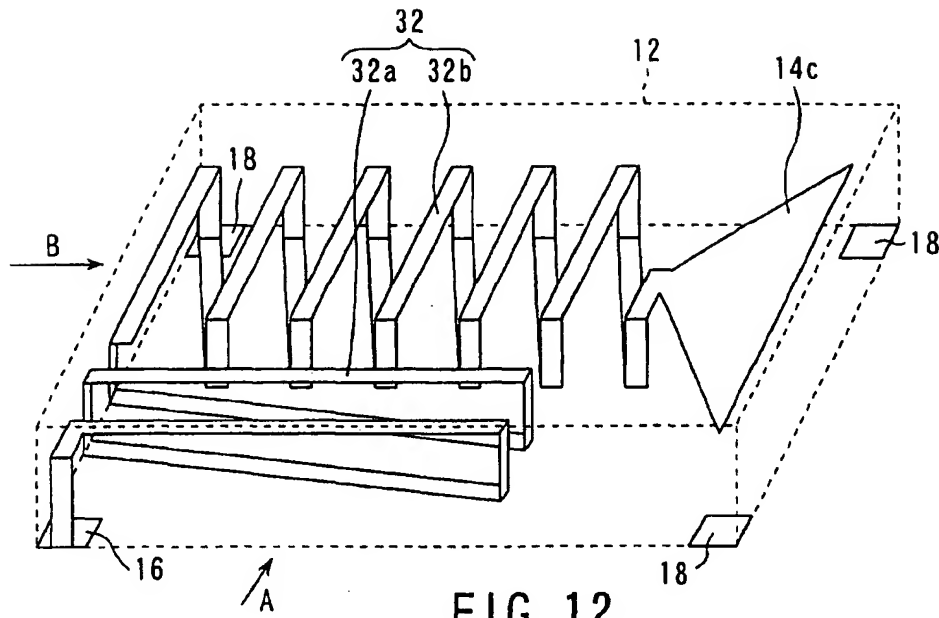


FIG. 11



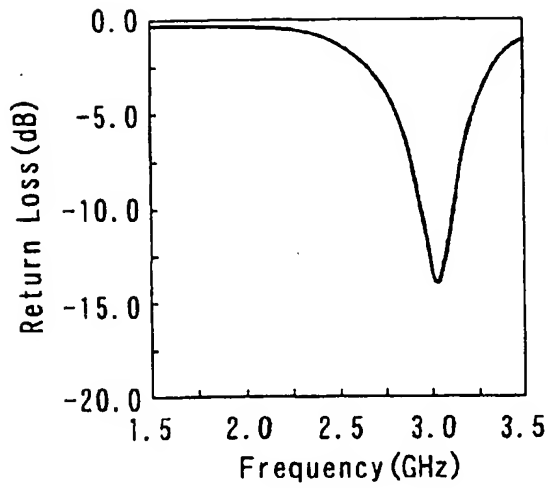


FIG. 14A

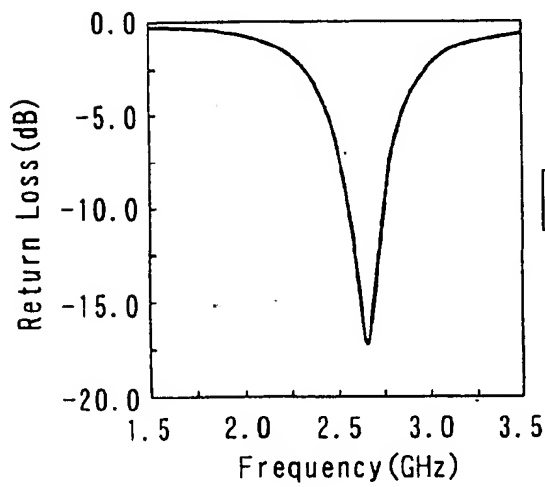


FIG. 14B

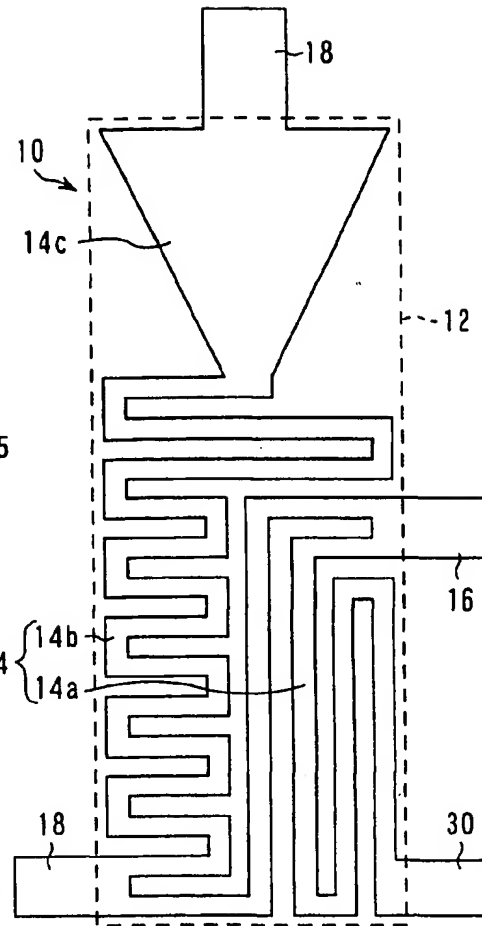


FIG. 15A

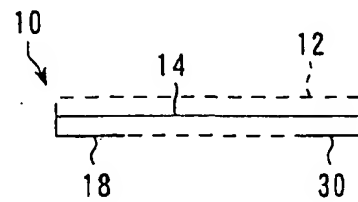


FIG. 15B

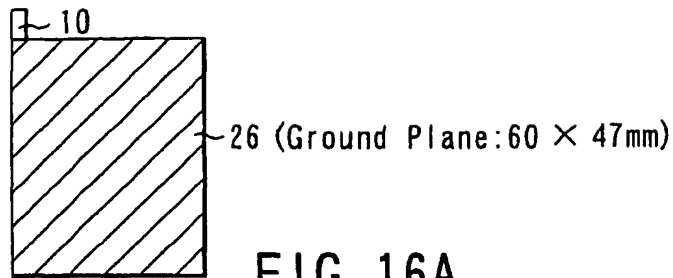


FIG. 16A



FIG. 16B

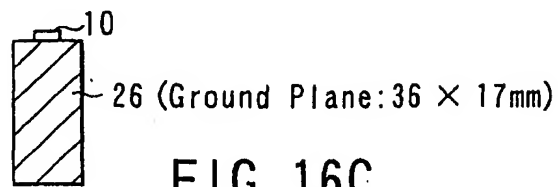


FIG. 16C

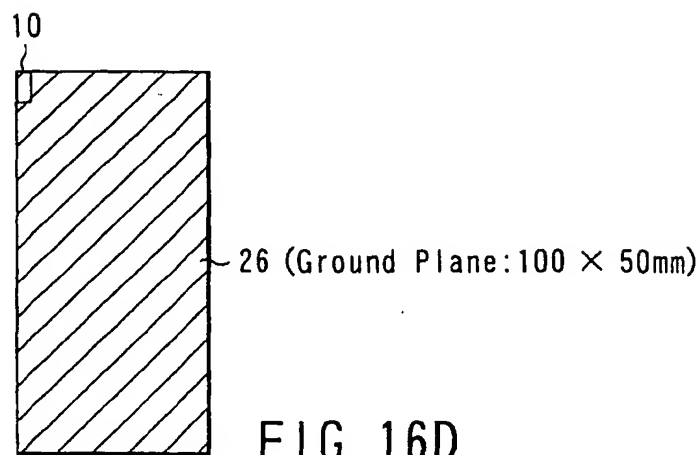
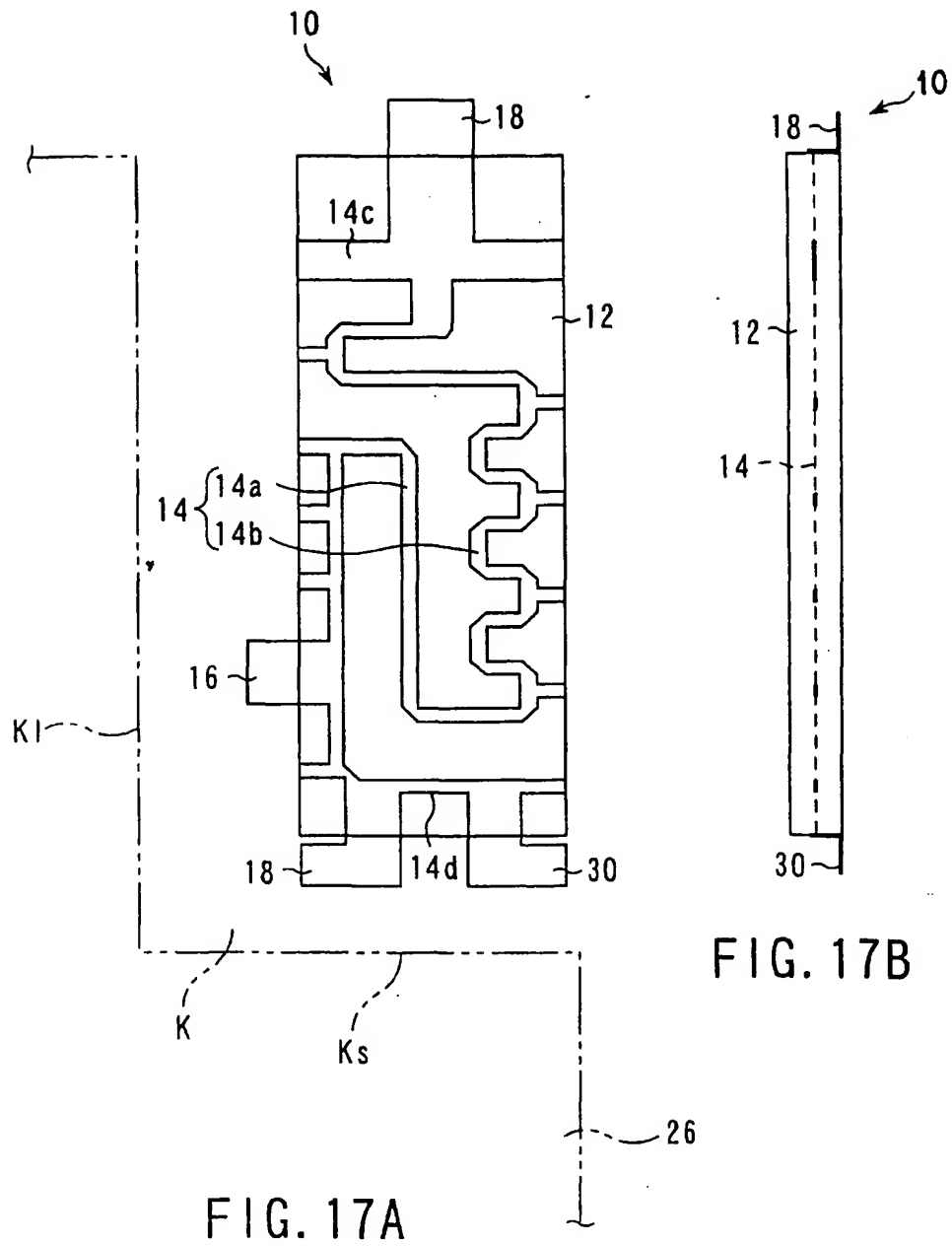


FIG. 16D





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| Place of search MUNICH | | Date of completion of the search 4 February 2002 | Examiner von Walter, S-U |
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